PROBLEM 3.1:

(a) There are 3 components $10\cos(800\pi t + \pi/4) = Re_{10}^{10}e^{j\pi/4}e^{j800\pi t}$ freg = 400 Hz $7\cos(1200\pi t - \pi/3) = Re\{7e^{-j\pi/3}e^{j1200\pi t}\}$ freq=600Hz -3cos(1600πt) = Re{3e^{jπ}e^{j1600πt}} freg=800Hz **↓**5e^{-jπ/4} 0 -400 -600 SPECTRUM (b) x(t) is periodic because there is a fundamental frequency f=200Hz that divides all 3 freqs. The period is the fundamental period = 1/200 sec (C) $5\cos(1000\pi t + \frac{\pi}{2}) = Re \{ 5e^{j\pi/2}e^{j1000\pi t} \}$ freg = 500Hz The spectrum will have two additional lines at f = -500 Hz and f = 500 Hz $5/2 e^{j\pi/2}$ component is <u>5</u>e^{jT/2} The dotted lines in the sketch above show where these two lines will be. Yes, y(t) is periodic. The fundamental frequency is now fo = 100 Hz because it has to divide into 500Hz as well as 400,600 and 800, The period is now too sec.

McClellan, Schafer and Yoder, Signal Processing First, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

PROBLEM 3.2:

(a) Read values from the graph:

$$x(t) = 4e^{j\pi/2}e^{-j2\pi(175)t} + 7e^{j\pi/3}e^{-j2\pi(50)t} + 11e^{j0t} + 4e^{-j\pi/2}e^{j2\pi(175)t} + 7e^{j\pi/3}e^{+j2\pi50t}$$
• Combine the positive is negative freqs:

$$x(t) = 8\cos(2\pi(175)t - \pi/2) + 14\cos(2\pi(50)t - \pi/3) + 11$$
(b) Cosine at $w = 2\pi(175)$ has $period = \frac{2\pi}{w} = \frac{1}{175}$
Cosine @ $w = 2\pi(50) \implies period = \frac{1}{50}$
11 is a constant $\implies freq = 0 \implies any period.$

$$\min period = \frac{2}{50} = \frac{1}{25} \sec.$$

(c)
$$y(t) = A \cos(\omega_0 t + \varphi)$$

 $= \operatorname{Re} \{ A \in i \varphi \in j \otimes t \}$
 $= \frac{1}{2} A e^{j\varphi} e^{j \otimes 0} t + \frac{1}{2} A e^{j\varphi} e^{-j \otimes t}$
 $Z_1 = \frac{1}{2} A e^{j\varphi} Z_1^* = \frac{1}{2} A e^{j\varphi} FREQ$

McClellan, Schafer and Yoder, Signal Processing First, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

PROBLEM 3.3:

(a)
$$X(t) = \sin^{3}(27\pi t)$$

 $= \left(\frac{1}{2j}e^{j27\pi t} - \frac{1}{2j}e^{j27\pi t}\right)^{3}$
 $= \frac{j}{8}\left(e^{j8\pi t} - 3e^{j54\pi t} - j^{27\pi t}\right)^{3}$
 $= \frac{j}{8}\left(e^{j8\pi t} - 3e^{j54\pi t} - e^{-j^{8\pi t}}\right)^{3}$
 $X(t) = \frac{1}{8}e^{j\pi/2}e^{j8\pi t} + \frac{3}{8}e^{j\pi/2}e^{j27\pi t} + \frac{3}{8}e^{j\pi/2}e^{-j27\pi t}$
 $+\frac{1}{8}e^{j\pi/2}e^{-j8\pi t}$
(b) $e^{\pm j8\pi t}$ has period $= \frac{2\pi}{8\pi} = \frac{2}{8\pi}$
 $e^{\pm j27\pi t}$ has period $= \frac{2\pi}{8\pi} = \frac{2}{8\pi}$
 $e^{\pm j27\pi t}$ has period $= \frac{2\pi}{27\pi} = \frac{2}{27\pi} = 3(\frac{2}{8\pi})$.
 \therefore period $= \frac{2}{27} \sec$.
(c) $\frac{3}{8}e^{j\pi/2}} = \frac{1}{2} \sec$.
(c) $\frac{3}{8}e^{j\pi/2}} = \frac{1}{2} \sec$.
 $\frac{1}{8}e^{j\pi/2}} = \frac{1}{2} \frac{1}{8}e^{j\pi/2}}{e^{j\pi/2}} = \frac{1}{2} \frac{1}{8}e^{j\pi/2}}{e^{j\pi/2}} = \frac{1}{8}e^{j\pi/2}}$

McClellan, Schafer and Yoder, Signal Processing First, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003



PROBLEM 3.4:

(a)
$$x(t) = \Re e \{ A e^{j2\pi(f_c - f_a)t} \} + \Re e \{ B e^{j2\pi(f_c + f_a)t} \}$$

$$= \Re e \{ (A e^{j2\pi f_a t} + B e^{j2\pi f_a t}) e^{j2\pi f_c t} \}$$
(b) $\bar{x}(t) = ((B+A)cos(2\pi f_a t) + j(B-A)sin(2\pi f_a t)) e^{j2\pi f_c t}$
 $\Re e \{ \bar{x}(t) \} = (B+A)cos(2\pi f_a t) cos(2\pi f_c t) - (B-A)sin(2\pi f_a t) sin(2\pi f_c t))$

$$\Longrightarrow C = B + A$$

$$D = -(B-A) = A - B$$
when $A = B = 1$, we get $C = 2$ and $D = 0$ so the
sine terms drop out.
(c) Want $D = 2$ and $C = 0$
 $B + A = 0$
 $A - B = 2$ $f \Rightarrow A = 1$ and $B = -1$
Return to the original expression for $x(t)$
 $x(t) = \frac{1}{2}e^{j2\pi(f_c - f_a)t} + \frac{1}{2}e^{-j2\pi(f_c - f_a)t} - \frac{1}{2}e^{j2\pi(f_c + f_a)t}$
 $-\frac{1}{2}e^{-j2\pi(f_c + f_a)t} + \frac{1}{2}e^{-j2\pi(f_c - f_a)t} + \frac{1}{2}e^{j\pi}$

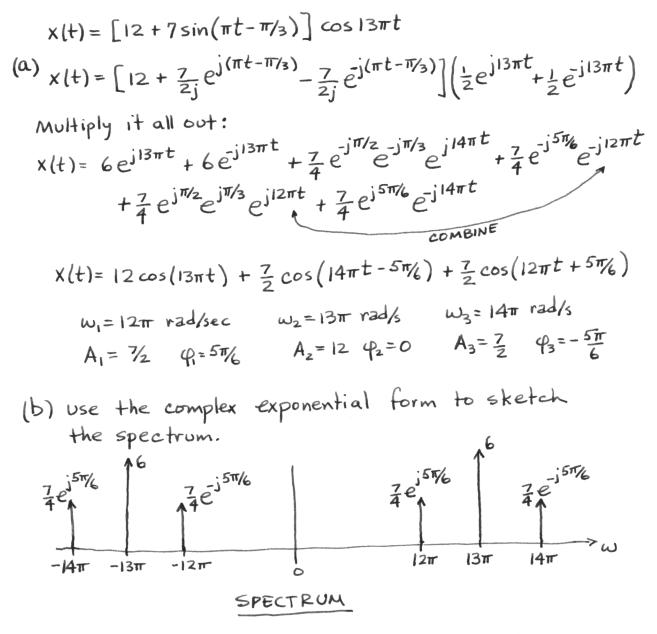
McClellan, Schafer and Yoder, *Signal Processing First*, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

PROBLEM 3.5:

Note: we use $\frac{1}{2}X_i$ because $\operatorname{Re}\{X_i\}=\frac{1}{2}X_i+\frac{1}{2}X_i^*$

McClellan, Schafer and Yoder, *Signal Processing First*, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

PROBLEM 3.6:



McClellan, Schafer and Yoder, Signal Processing First, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

PROBLEM 3.7:

$$X(t) = \cos[(\omega_1 + \omega_2)t] + \cos[(\omega_2 - \omega_1)t]$$
(a) If $x(t)$ is to be periodic, we need to find
a fundamental frequency ω_0 that divides
both $(\omega_1 + \omega_2)$ and $(\omega_2 - \omega_1)$. That is
 $\omega_1 + \omega_2 = l_1 \omega_0$ where $l_1 \notin l_2$ are
 $\omega_2 - \omega_1 = l_2 \omega_0$ integers. Also, $l_2 < l_1$
 $\implies \frac{\omega_2 - \omega_1}{\omega_2 + \omega_1} = \frac{l_2}{l_1} < 1$
Thus, the ratio of the sum and difference
frequencies must be a rational number.
(b) Now solve for ω_1/ω_2 and show that it
must be rational.
 $(\omega_2 - \omega_1) l_1 = (\omega_2 + \omega_1) l_2$
 $\omega_2(l_1 - l_2) = \omega_1(l_1 + l_2)$
 $\implies \frac{\omega_1}{\omega_2} = \frac{l_1 - l_2}{m_1} < 1$ where $m_1 \notin m_2$
are integers.
From above:
 $2\omega_2 = (l_1 + l_2)\omega_0$
 $\omega_1 = \frac{1}{2}(l_1 - l_2)\omega_0$
 $\implies \omega_1 \notin \omega_2$ are integer multiples of $\frac{1}{2}\omega_0$

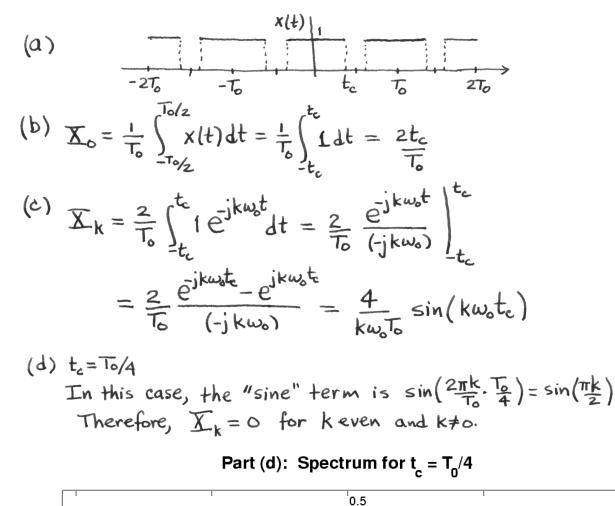
McClellan, Schafer and Yoder, Signal Processing First, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

PROBLEM 3.8:

x(t) has four components:
2 =
$$2e^{j^0}$$
 \implies freq=0, period = anything
 $4\cos(40\pi t - \pi/s) = Re\{4e^{j\pi/s}e^{j40\pi t}\}$ freq=20Hz
period=/20sec.
 $3\sin(60\pi t) = 3\cos(60\pi t - \pi/2)$
 $= Re\{3e^{j\pi/s}e^{j120\pi t}\}$ freq=30Hz
period = 1/30sec.
 $4\cos(120\pi t - \pi/3) = Re\{4e^{j\pi/s}e^{j120\pi t}\}$ freq=60Hz
period = 1/60sec.
(a) The fundamental period is the smalkest time that is
 $exactly divisible by 1/20, 1/30 and 1/60$.
 $\implies T_0 = 1/0 sec$ because $T_0 = 2(1/20) = 3(1/50) = 6(1/60)$
 $\implies T_0 = 10 \text{ Hz} \implies \omega_0 = 2\pi f_0 = 20\pi \text{ rad/sec}$.
 $X_0 = 2, X_2 = 4e^{j\pi/s}, X_3 = 3e^{j\pi/s}, X_6 = 4e^{j\pi/3}$
(b) $2e^{j\pi/3}, 3e^{j\pi/s}, 2e^{j\pi/s}, 1/2, 2e^{j\pi/s}, 3e^{j\pi/s}, 1/2, 4e^{j\pi/s}$
 $Re\{10e^{j\pi/s}e^{j50\pi t}\}$ freq=25Hz
period=1/25 sec
The new period must be divisible by 1/0 $\frac{1}{2}/25$
 $\implies T_0 = 1/5$ because $T_0 = 5(1/25)$ and $2(1/0)$
 $\implies T_0 = 1/6 sec$ $T_0 = 5(1/25)$ and $2(1/0)$
 $\implies F_0 = 5\text{ Hz} \implies \omega_0 = 2\pi f_0 = 10\pi \text{ rad/sec}$.
We must re-index the Xi because ω_0 is lower.
 $Y_0 = 2, Y_4 = X_2, Y_5 = 10e^{j\pi/6}, Y_6 = X_3, Y_{12} = X_6$
There will be one new pair of lines at $f = 125\text{ Hz}$

McClellan, Schafer and Yoder, Signal Processing First, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

PROBLEM 3.9:



0.32

-0.11

0.32

0 frequency (Hz)

-0.11

0.064

-500



-0.045

Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

0.035

Magnitude 0.2

> 0 -1000

> > This page should not be copied or electronically transmitted unless prior written permission has been obtained from the authors. December 30, 2003

-0.045

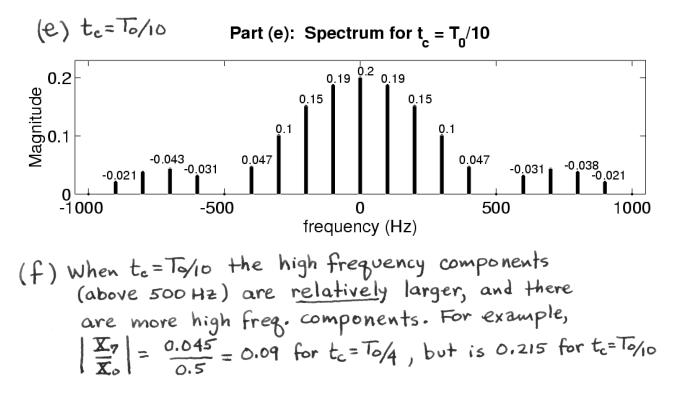
0.035

1000

0.064

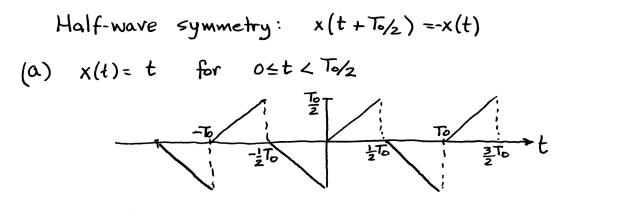
500

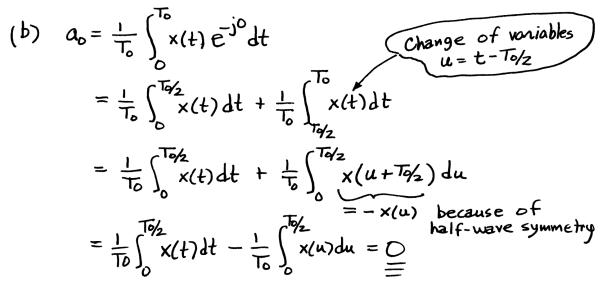
PROBLEM 3.9 (more):



McClellan, Schafer and Yoder, Signal Processing First, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

PROBLEM 3.10:





(c) If k is even, then k= 2l

$$a_{k} = \frac{1}{T_{0}} \int_{0}^{T_{0}/2} x(t) e^{j(2\pi/T_{0})2lt} dt + \frac{1}{T_{0}} \int_{T_{0}/2}^{T_{0}} x(t) e^{j(2\pi/T_{0})2lt} dt$$
Make the same change of variables: $u = t - T_{0}/2$.

$$a_{k} = \frac{1}{T_{0}} \int_{0}^{T_{0}/2} x(t) e^{-j(2\pi/T_{0})2lt} dt + \frac{1}{T_{0}} \int_{0}^{T_{0}/2} x(u + \frac{1}{2}) e^{-j(2\pi/T_{0})2l(u + T_{0}/2)} du$$

$$e^{-j(2\pi/T_{0})2lu} e^{-j(2\pi/T_{0})2lT_{0}/2} = e^{-j(2\pi/T_{0})2lu} e^{-j2\pi l}$$

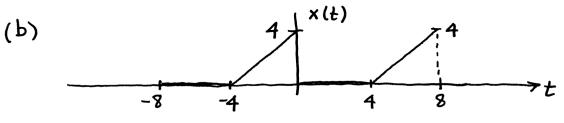
$$a_{k} = \frac{1}{T_{0}} \int_{0}^{T_{0}/2} x(t) e^{-j(2\pi/T_{0})2lt} dt - \frac{1}{T_{0}} \int_{0}^{T_{0}/2} x(u) e^{-j(2\pi/T_{0})2lu} du$$

$$\Rightarrow a_{k} = 0 \text{ for } k \text{ even}.$$

McClellan, Schafer and Yoder, Signal Processing First, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

PROBLEM 3.11:

$$\begin{aligned} \Omega_{k} &= \frac{1}{8} \int_{-4}^{0} (4+t) e^{-j(2\pi/8)kt} dt \\ &= \frac{1}{7} \int_{-\frac{1}{2}}^{\frac{1}{2}} x(t) e^{-j(2\pi/7)kt} dt \\ (a) \quad T &= 8 \text{ secs } \frac{1}{8} x(t) = \begin{cases} (4+t) & \text{for } -4 \leq t \leq 0 \\ 0 & \text{for } 0 < t < 4 \end{cases} \end{aligned}$$



(c)

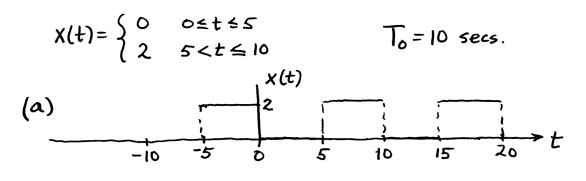
$$a_{0} = \frac{1}{8} \int_{-4}^{0} (t+4) e^{-j0} dt = \frac{1}{8} (\frac{t^{2}}{2} + 4t) \Big|_{-4}^{0}$$

$$= 0 - \frac{1}{8} (\frac{16}{2} - 16) = 1$$
or, $a_{0} = \frac{1}{8} \times \text{Area-in-one-period}$

$$= \frac{1}{8} \times (\frac{1}{2} \times 4 \times 4) = 1$$

McClellan, Schafer and Yoder, Signal Processing First, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

PROBLEM 3.12:



(b)
$$a_0 = \frac{1}{T_0} \times \text{Area} = \frac{1}{10} \times (5 \times 2) = 1$$

(c)
$$a_{1} = \frac{1}{10} \int_{5}^{10} 2e^{-j(2\pi/6)t} dt$$

 $= \frac{1}{5} \frac{e^{-j\pi t/5}}{(-j\pi/5)} \Big|_{5}^{10} = \frac{e^{-j^{2}\pi} - e^{-j\pi}}{-j\pi} = \frac{1 - (-1)}{-j\pi} = \frac{2j}{\pi}$
(d) $y(t) = 1 + x(t) = 1 + \sum_{k=0}^{\infty} a_{k} e^{jw_{k}kt}$

$$= (1+a_0) + \sum_{k\neq 0}^{r} a_k e^{j\omega_0 kt}$$

$$\implies b_0 = 1+a_0 \quad \text{and} \quad b_1 = a_1$$

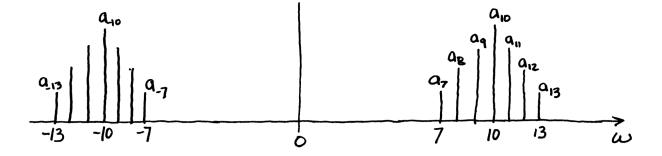
McClellan, Schafer and Yoder, Signal Processing First, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

PROBLEM 3.13:

$$\begin{aligned} x(t) &= \sin(10t) \left(\sum_{k=-3}^{3} \frac{1}{1+jk} e^{jkt} \right) \\ &= \left(\frac{1}{2j} e^{j10t} - \frac{1}{2j} e^{-j10t} \right) \left(\sum_{k=-3}^{3} \frac{1}{1+jk} e^{jkt} \right) \\ &= \sum_{k=-3}^{3} \frac{1}{2j-2k} e^{j(k+10)t} + \sum_{k=-3}^{3} \frac{1}{-2j+2k} e^{j(k-10)t} \\ &\quad \text{change index} \\ &\quad t = k+10 \end{aligned}$$

Note: Wo = 1 rad/s for this signal. The sums above contain the 7th through 13th harmonics.

$$a_{k} = \begin{cases} \frac{1}{2j-2k+20} & k=7,8,9,10,11,12,13\\ \frac{1}{-2j+2k+20} & k=-7,-8,-9,-10,-11,-12,-13\\ 0 & elsewhere \end{cases}$$



McClellan, Schafer and Yoder, Signal Processing First, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

PROBLEM 3.14:

(a)
$$y(t) = Ax(t) = A \cdot \sum_{k} a_{k} e^{jk\omega_{0}t}$$

= $\sum_{k} (Aa_{k}) e^{jk\omega_{0}t} \implies b_{k} = Aa_{k}$

(b)
$$y(t) = x(t-t_d)$$

$$= \sum_{\kappa} a_{\kappa} e^{j\omega_{\delta}k(t-t_d)}$$

$$= \sum_{\kappa} (a_{\kappa} e^{-j\omega_{\delta}kt_d}) e^{j\omega_{\delta}kt}$$

$$\implies b_{\kappa} = a_{\kappa} e^{-jk\omega_{\delta}t_d}$$

McClellan, Schafer and Yoder, Signal Processing First, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

PROBLEM 3.15:

(a)

$$\begin{aligned}
\mathbf{a}_{\mathsf{K}} &= \frac{1}{\mathsf{T}_{\mathsf{o}}} \int_{-\mathsf{T}_{\mathsf{o}/\mathsf{A}}}^{\mathsf{T}_{\mathsf{o}/\mathsf{A}}} \mathbf{1} e^{-j(2\pi/\mathsf{T}_{\mathsf{o}})\mathsf{k}\mathsf{t}} \, \mathsf{d}\mathsf{t} \\
&= \frac{1}{\mathsf{T}_{\mathsf{o}}} \left(\frac{e^{-j(2\pi/\mathsf{T}_{\mathsf{o}})\mathsf{k}\mathsf{t}}}{(-j\,\mathsf{k})(2\pi/\mathsf{T}_{\mathsf{o}})} \right)_{-\mathsf{T}_{\mathsf{o}/\mathsf{A}}}^{\mathsf{T}_{\mathsf{o}/\mathsf{A}}} \\
&= \frac{e^{-j(2\pi/\mathsf{T}_{\mathsf{o}})\mathsf{k}\mathsf{T}_{\mathsf{o}/\mathsf{A}}}}{-j\,2\pi/\mathsf{k}} - e^{+j(2\pi/\mathsf{T}_{\mathsf{o}})\mathsf{k}\mathsf{T}_{\mathsf{o}/\mathsf{A}}} \\
&= \frac{e^{-j(2\pi/\mathsf{T}_{\mathsf{o}})\mathsf{k}\mathsf{T}_{\mathsf{o}/\mathsf{A}}}}{-j\,2\pi/\mathsf{k}} = \frac{-j\pi/\mathsf{k}/2}{-j\,2\pi/\mathsf{k}}} = \frac{-j\pi/\mathsf{k}/2}{\pi/\mathsf{k}}} \\
\end{aligned}$$

Note: a= 1/2

(b)
$$y(t) = 2x(t - T_0/2)$$

If $y(t) = \sum_{k} b_{k} e^{jk\omega_{b}t} \stackrel{!}{\underset{k}{=}} x(t) = \sum_{k} a_{k} e^{jk\omega_{b}t}$
Then $b_{k} = 2a_{k} e^{-jk\omega_{b}T_{0}/2}$
Since $\omega_{o} = 2\pi/T_{0}$ the exponent is $-jk\pi$
 $\Rightarrow b_{k} = (2e^{-j\pi k})a_{k} = (2(-1)^{k})a_{k}$
 $\stackrel{!}{\underset{k}{=}} b_{k} = 2(-1)^{k} \frac{\sin(\pi k/2)}{\pi k}$

McClellan, Schafer and Yoder, Signal Processing First, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

PROBLEM 3.16:

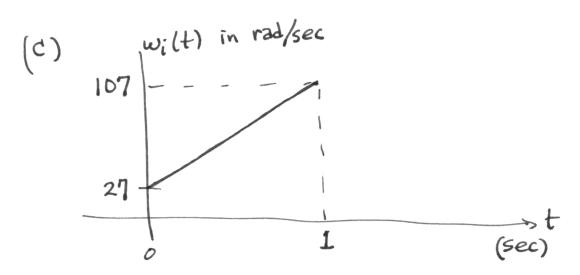
(a) The formula is (440) 2 ^{K/12} where k is the number of keys above or below A-440.					
key	k	freg(Hz)	key	k	freg (HZ)
С С#	-9 -8	261.6 277.2	F [⊭] G	-3 -2	370.0 392.0
D E	-7	293.7	Ч G#	-1	415.3
E		311.1	A	0	440
F	-4	329.6 349.2	B ^b B	1	466.2
			C	2	523.3
(b) $f = (440) 2^{(n-49)/12}$ key number for A-440					
(c) D-F#-A has three frequencies					
$Ae^{j\varphi_3}$ $Ae^{j\varphi_2}$ $Ae^{j\varphi_1}$ $Ae^{j\varphi_1}$ $Ae^{j\varphi_2}$ $Ae^{j\varphi_3}$					
-440 -370 -293.7 "293.7 370 440 f(Hz)					
SPECTRUM of TYPICAL D-Major CHORD					
Assuming amplitude of each note is the same					

McClellan, Schafer and Yoder, Signal Processing First, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

PROBLEM 3.17:

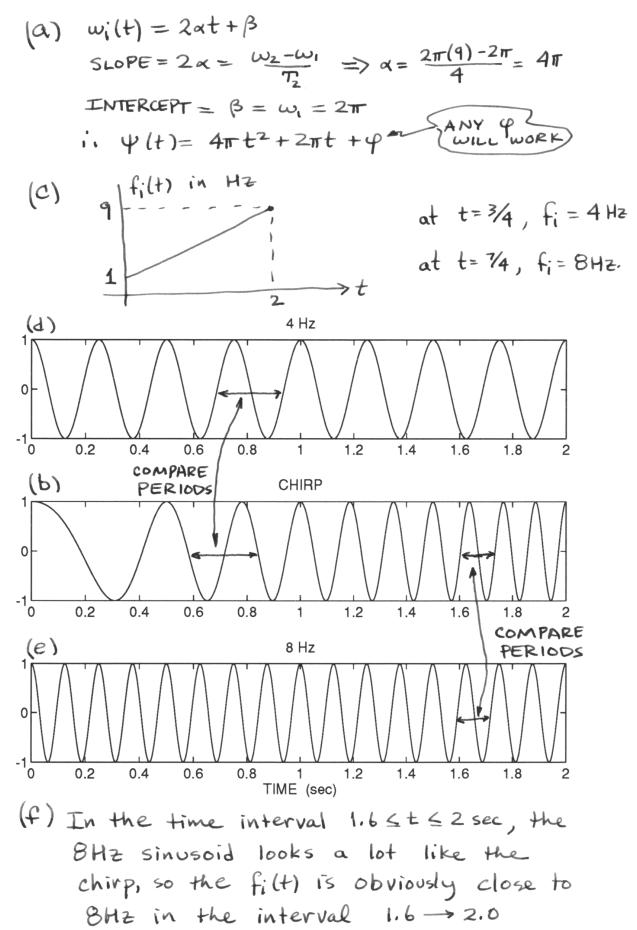
(a)
$$w_i(t) = \frac{d}{dt} \psi(t) = 2\alpha t + \beta$$

 $w_i = w_i(t)|_{t=0} = \beta$
 $w_2 = w_i(t)|_{t=T_2} = 2\alpha T_2 + \beta$
(b) $\psi(t) = 40t^2 + 27t + 13$
 $w_i(t) = \frac{d}{dt} \psi = 80t + 27$



McClellan, Schafer and Yoder, Signal Processing First, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

PROBLEM 3.18:



McClellan, Schafer and Yoder, Signal Processing First, ISBN 0-13-065562-7. Pearson Prentice Hall, Inc. Upper Saddle River, NJ 07458. © 2003

Characterize each time signal:

- (a) 6 periods from t=-2 to $t=+3 \implies T=\frac{5}{6}$ sec DC value = 2 $t_m > 0 \implies q < 0$
- (b) 3 periods from t=0 to t=2 \implies T= $\frac{2}{3}$ sec DC value = 0 $\varphi = \pi$
- (c) 6 periods from t=-3 to $t=2 \implies T=\frac{2}{5}$ sec DC value = 2 $t_m < 0 \implies q > 0$
- (d) Period $\approx 3\frac{1}{3} = \frac{19}{3}$ secs. Two frequencies DC value = 0
- (e) 2 periods from t=-2 to t=3 => T=2.5 secs Two frequencies. DC value =0
- (1) $\omega_0 = 2\pi (1.2) \implies T = \frac{1}{1.2} = \frac{5}{6} \sec \varphi = 0.5\pi > 0$ DC = 2
- (2) $w_0 = 2\pi (0.3)$ because 0.3 divides 0.6 \$ 1.5 $\Rightarrow T = \frac{1}{0.3} = \frac{10}{3}$ secs. DC = 0
- (3) Like (1). $T = \frac{5}{6} \sec$. DC = 2 But $\varphi = -0.25\pi < 0$
- (4) $\omega_0 = 2\pi (0.4) \implies T = \frac{1}{0.4} = 2.5 \text{ secs}$ DC = 0

(5)
$$\varphi = \pi$$
 $T = \frac{1}{1.5} = \frac{2}{3} \sec \theta$

Thus, the match is

$$(a) \leftrightarrow (3)$$
 $(c) \leftrightarrow (1)$ $(e) \leftrightarrow (4)$
 $(b) \leftrightarrow (5)$ $(d) \leftrightarrow (2)$