



PROBLEM 6.1:

$$x[n] = e^{-j\pi/2} e^{j0.4\pi n}$$

$$e^{-j\pi/2} = -j$$

$$y[n] = x[n] - x[n-1]$$

$$= e^{-j\pi/2} e^{j0.4\pi n} - e^{-j\pi/2} e^{j0.4\pi(n-1)}$$

$$= e^{-j\pi/2} e^{j0.4\pi n} (1 - e^{-j0.4\pi})$$

ANGLE = 54°
or 0.942 rads

$$= 1.176 e^{j0.3\pi}$$

$$= 1.176 e^{-j0.2\pi} e^{j0.4\pi n}$$

$$A = 1.176 \quad \varphi = -0.2\pi = -0.628 \text{ rads}, \alpha = -36^\circ$$



PROBLEM 6.2:

$$y[n] = (x[n])^2$$

(a) $x[n] = Ae^{j\varphi} e^{j\hat{\omega}n}$

$$y[n] = (Ae^{j\varphi} e^{j\hat{\omega}n})^2 = A^2 e^{j2\varphi} e^{j2\hat{\omega}n}$$

(b) NO.

The output cannot be written as

$$y[n] = \chi(\hat{\omega}) Ae^{j\varphi} e^{j\hat{\omega}n}$$

because the frequency has changed

The new freq. is $2\hat{\omega}$

PROBLEM 6.3:



$$y[n] = x[-n]$$

$$(a) x[n] = Ae^{j\varphi} e^{j\hat{\omega}n}$$

$$y[n] = Ae^{j\varphi} e^{j\hat{\omega}(-n)} = Ae^{j\varphi} e^{-j\hat{\omega}n}$$

(b) NO.

The output cannot be written as

$$y[n] = \mathcal{H}(\hat{\omega}) Ae^{j\varphi} e^{j\hat{\omega}n}$$

because the frequency has changed from $+\hat{\omega}$ to $-\hat{\omega}$.

PROBLEM 6.4:



$$y[n] = 2x[n] - 3x[n-1] + 2x[n-2]$$

$$(a) \quad H(\hat{\omega}) = \sum_{k=0}^2 b_k e^{-j\hat{\omega}k} = 2 - 3e^{-j\hat{\omega}} + 2e^{-j2\hat{\omega}}$$

Simplify with symmetry:

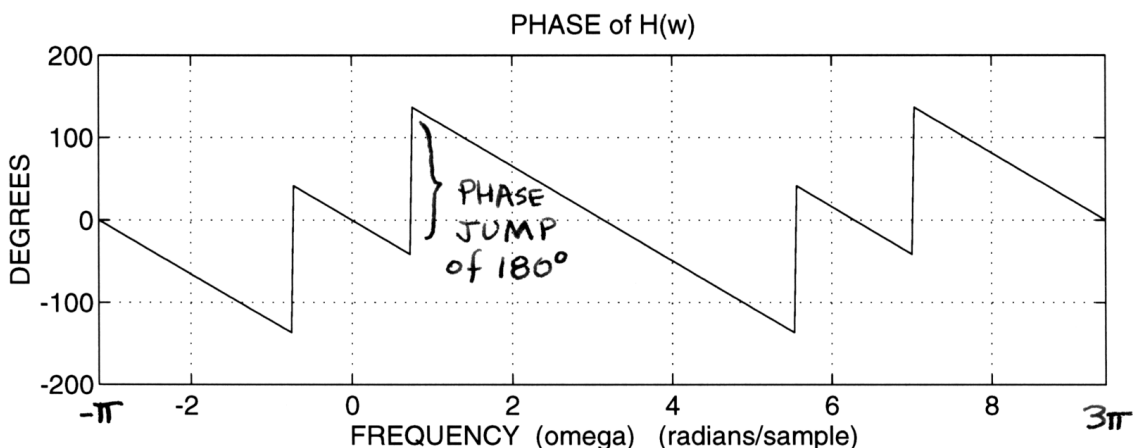
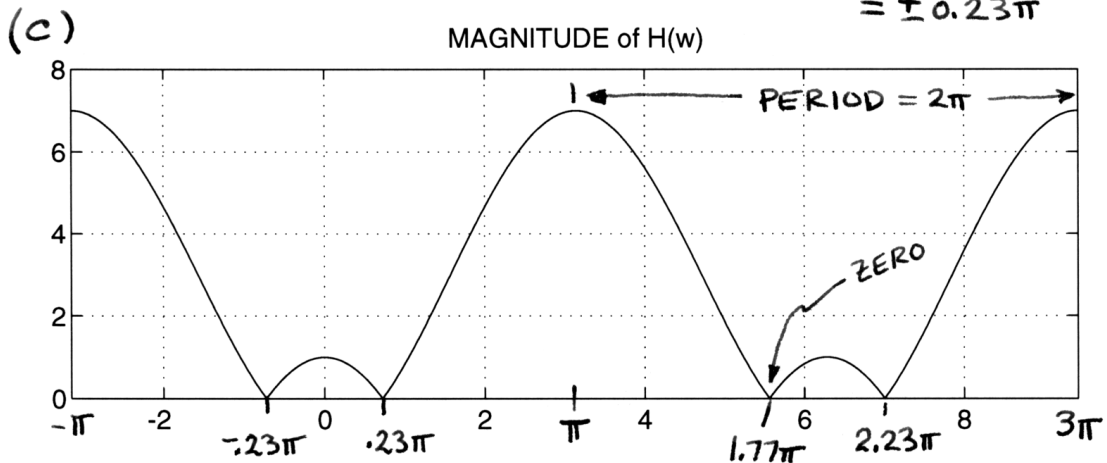
$$H(\hat{\omega}) = e^{-j\hat{\omega}} \{ 2e^{j\hat{\omega}} - 3 + 2e^{-j\hat{\omega}} \}$$

$$= \underbrace{(4 \cos \hat{\omega} - 3)}_{\text{PHASE}} e^{-j\hat{\omega}}$$

THIS TERM IS MAG, EXCEPT FOR A POSSIBLE MINUS SIGN

(b) $\cos \hat{\omega}$ has PERIOD = $2\pi \Rightarrow H(\hat{\omega})$ has period = 2π

(d) OUTPUT IS ZERO WHEN $H(\hat{\omega}) = 0$
 \Rightarrow SOLVE $4 \cos \hat{\omega} - 3 = 0 \Rightarrow \hat{\omega} = \cos^{-1}(3/4)$
 $= \pm 0.23\pi$





PROBLEM 6.4 (more):

(e) $\mathcal{X}(\hat{\omega})$ at $\hat{\omega} = \pi/13$ is $\mathcal{X}(\pi/13) = 0.8838 e^{-j0.077\pi}$

Since the freq. response alters mag & phase of the input, we can get output via:

$$x[n] = \sin\left(\frac{\pi}{13}n\right) = \cos\left(\frac{\pi}{13}n - \frac{\pi}{2}\right).$$

$$\begin{aligned} \Rightarrow y[n] &= 0.8838 \cos\left(\frac{\pi}{13}n - 0.5\pi - 0.077\pi\right) \\ &= 0.8838 \cos\left(\frac{\pi}{13}n - 0.577\pi\right) \quad \text{for all } n \end{aligned}$$

Another approach which uses linearity:

$$\sin\left(\frac{\pi}{13}n\right) = \frac{1}{2j} e^{j\pi n/13} - \frac{1}{2j} e^{-j\pi n/13}$$

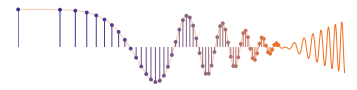
$$\Rightarrow y[n] = \mathcal{X}\left(\frac{\pi}{13}\right) \frac{1}{2j} e^{j\pi n/13} - \mathcal{X}\left(-\frac{\pi}{13}\right) \frac{1}{2j} e^{-j\pi n/13}$$

$$= \frac{1}{2j} \left\{ 0.8838 e^{j(\pi/13 - 0.077\pi)} - 0.8838 e^{-j(\pi/13 - 0.077\pi)} \right\}$$

$$= 0.8838 \sin\left(\frac{\pi}{13}n - 0.077\pi\right)$$

$$\text{or } y[n] = 0.8838 \cos\left(\frac{\pi}{13}n - 0.577\pi\right).$$

NOTE: THIS METHODS TRACKS THE POSITIVE AND NEGATIVE FREQUENCY COMPONENTS THROUGH THE SYSTEM SEPARATELY.



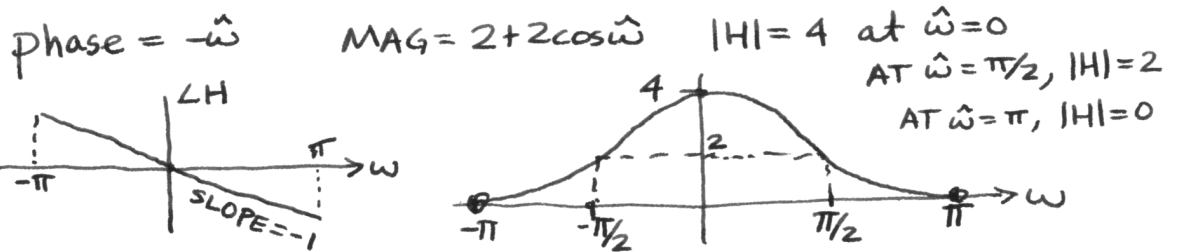
PROBLEM 6.5:

$$y[n] = x[n] + 2x[n-1] + x[n-2]$$

(a) use filter coeffs: $\{b_k\} = \{1, 2, 1\}$

$$\mathcal{H}(\hat{\omega}) = 1 + 2e^{-j\hat{\omega}} + e^{-j2\hat{\omega}}$$

(b) $\mathcal{H}(\hat{\omega}) = e^{-j\hat{\omega}}(e^{j\hat{\omega}} + 2 + e^{-j\hat{\omega}}) = e^{-j\hat{\omega}}(2 + 2\cos\hat{\omega})$



(c) $x[n] = 10 + 4\cos(\frac{\pi}{2}n + \frac{\pi}{4})$

$$= 10 + 2e^{j\pi/4}e^{j\pi/2n} + 2e^{-j\pi/4}e^{-j\pi/2n}$$

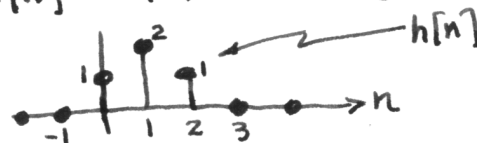
$$y[n] = 10\mathcal{H}(0) + \mathcal{H}(\pi/2)2e^{j\pi/4}e^{j\pi/2n} + 2\mathcal{H}(-\pi/2)e^{-j\pi/4}e^{-j\pi/2n}$$

$$\mathcal{H}(0) = 4e^{j0} \quad \mathcal{H}(\pi/2) = e^{-j\pi/2}(2) \quad \mathcal{H}(-\pi/2) = 2e^{j\pi/2}$$

$$\Rightarrow y[n] = 40 + 4e^{-j\pi/2}e^{j\pi/4}e^{j\pi/2n} + 4e^{j\pi/2}e^{-j\pi/4}e^{-j\pi/2n}$$

$$= 40 + 8\cos(\frac{\pi}{2}n - \frac{\pi}{4})$$

(d) $x[n] = \delta[n] \Rightarrow y[n] = \delta[n] + 2\delta[n-1] + \delta[n-2]$



(e) $x[n] = u[n]$

$$y[n] = u[n] + 2u[n-1] + u[n-2]$$

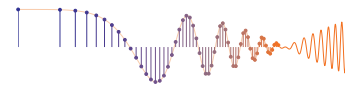
$$y[n] = 0 \text{ for } n < 0$$

$$y[0] = u[0] + 2u[-1] + u[-2] = 1 + 0 + 0 = 1$$

$$y[1] = u[1] + 2u[0] + u[-1] = 1 + 2 + 0 = 3$$

$$y[2] = u[2] + 2u[1] + u[0] = 1 + 2 + 1 = 4$$

$$y[n] = 4 \text{ for } n \geq 2.$$



PROBLEM 6.6:

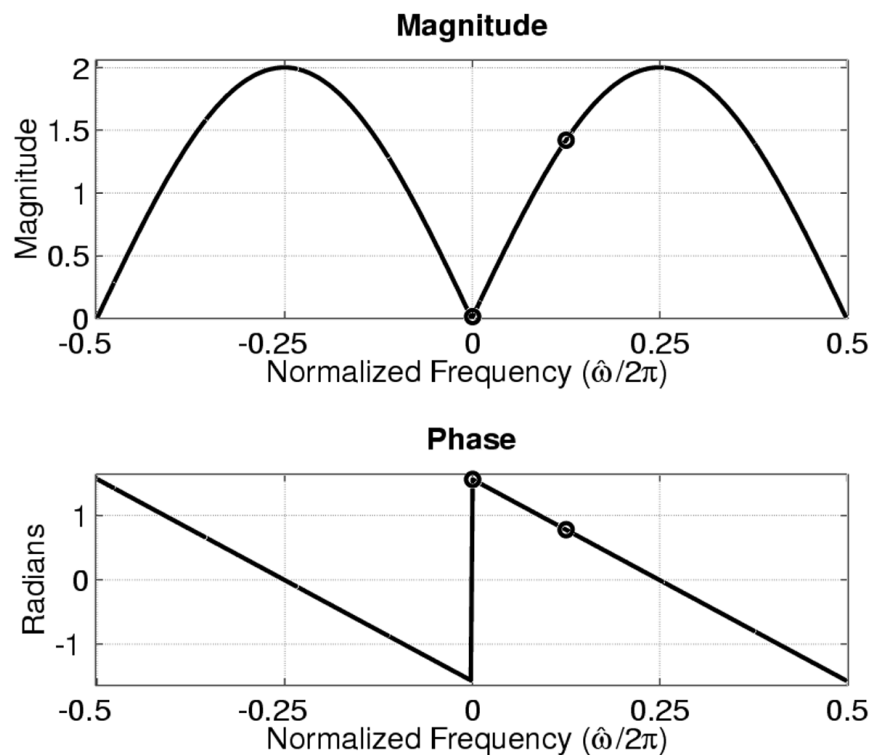
(a) $y[n] = x[n] - x[n-2]$

The filter coefficients are $\{b_k\} = \{1, 0, -1\}$

$$H(\hat{\omega}) = \sum_{k=0}^2 b_k e^{-j\hat{\omega}k} = 1 - e^{-j2\hat{\omega}}$$

$$H(\hat{\omega}) = e^{-j\hat{\omega}}(e^{j\hat{\omega}} - e^{-j\hat{\omega}}) = 2e^{j\pi/2} e^{-j\hat{\omega}} \sin \hat{\omega}$$

(b) MATLAB plots



(c) $x[n] = 4 + \cos(\frac{\pi}{4}n - \frac{\pi}{4})$

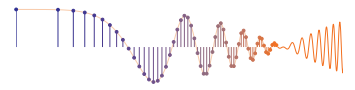
Need $H(0)$ Need $H(\frac{\pi}{4})$

$$H(0) = 2e^{j\pi/2} e^{-j0} \sin(0) = 0$$

$$H(\frac{\pi}{4}) = 2e^{j\pi/2} e^{-j\pi/4} \sin(\pi/4) = \sqrt{2} e^{j\pi/4}$$

PHASE CHANGE

$$y[n] = 0 + \sqrt{2} \cos(\frac{\pi}{4}n - \frac{\pi}{4} + \frac{\pi}{4}) = \sqrt{2} \cos(\frac{\pi}{4}n) \text{ for all } n$$



PROBLEM 6.6 (more):

(d) $x_1[n] = x[n]u[n]$, so $x_1[n] = 0$ for $n < 0$

The FIR filter is

$$y[n] = x[n] - x[n-2]$$

Once $(n-2) \geq 0$, the same numbers are involved in the calculation, so $y_1[n] = y[n]$ for $n \geq 2$. When $n < 2$, then $x_1[n-2] = 0$ and $x_1[n-2] \neq x[n-2]$, so $y_1[n] \neq y[n]$ for $n < 2$.

Here are the actual values for $y_1[n]$:

$$y_1[n] = 0 \text{ for } n < 0$$

$$y_1[0] = x_1[0] - x_1[-2] = x_1[0] = 4 + \sqrt{2}/2$$

$$y_1[1] = x_1[1] - x_1[-1] = x_1[1] = 4 + 1 = 5$$

$$y_1[2] = x_1[2] - x_1[0] = 4 + \sqrt{2}/2 - (4 + \sqrt{2}/2) = 0$$

$$y_1[n] = \sqrt{2} \cos\left(\frac{\pi}{4}n\right) \text{ for } n \geq 2.$$



PROBLEM 6.7:

(a) $H(e^{j\hat{\omega}}) = 1 + 2e^{-j3\hat{\omega}}$

Solution: Use the fact that the frequency response for $\delta[n - n_0]$ is $H(e^{j\hat{\omega}}) = e^{-j3\hat{\omega}}$.

$$h[n] = \delta[n] + 2\delta[n - 3]$$

(b) $H(e^{j\hat{\omega}}) = 2e^{-j3\hat{\omega}} \cos(\hat{\omega})$

Solution: Use the inverse Euler formula to write the frequency response in terms of complex exponentials.

$$H(e^{j\hat{\omega}}) = 2e^{-j3\hat{\omega}} \cos(\hat{\omega}) = e^{-j3\hat{\omega}} (e^{j\hat{\omega}} + e^{-j\hat{\omega}})$$

$$H(e^{j\hat{\omega}}) = e^{-j2\hat{\omega}} + e^{-j4\hat{\omega}}$$

$$\Rightarrow h[n] = \delta[n - 2] + \delta[n - 4]$$

(c) $H(e^{j\hat{\omega}}) = e^{-j4.5\hat{\omega}} \frac{\sin(5\hat{\omega})}{\sin(\hat{\omega}/2)}$

Solution: Use the fact that the frequency response for an L -point running sum filter is:

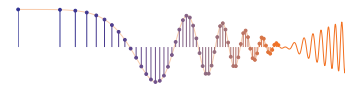
$$H_L(e^{j\hat{\omega}}) = e^{-j\hat{\omega}(L-1)/2} \frac{\sin(L\hat{\omega}/2)}{\sin(\hat{\omega}/2)}$$

Thus, we see that $L/2 = 5$, or $L = 10$, and we can rewrite the frequency response as

$$H(e^{j\hat{\omega}}) = e^{-j\hat{\omega}(10-1)/2} \frac{\sin(10\hat{\omega}/2)}{\sin(\hat{\omega}/2)}$$

because $(L - 1)/2$ is equal to 4.5 when $L = 10$. Having made these identifications in the formula for $H(e^{j\hat{\omega}})$, we get the impulse response of the 10-point running-sum filter:

$$\begin{aligned} h[n] &= u[n] - u[n - 10] \\ &= \delta[n] + \delta[n - 1] + \delta[n - 2] + \delta[n - 3] + \delta[n - 4] \dots \\ &\quad + \delta[n - 5] + \delta[n - 6] + \delta[n - 7] + \delta[n - 8] + \delta[n - 9] \end{aligned}$$



PROBLEM 6.8:

$$\begin{aligned}
 (a) \quad H(\hat{\omega}) &= (1 + e^{-j\hat{\omega}}) \underbrace{(1 - 2\cos(2\pi/3)e^{-j\hat{\omega}} + e^{-j2\hat{\omega}})}_{=2(1/2)=1} \\
 &= (1 + e^{-j\hat{\omega}})(1 - e^{-j\hat{\omega}} + e^{-j2\hat{\omega}}) \\
 &= 1 - e^{-j\hat{\omega}} + e^{-j2\hat{\omega}} + e^{-j\hat{\omega}} - e^{-j2\hat{\omega}} + e^{-j3\hat{\omega}} \\
 &= 1 + e^{-j3\hat{\omega}}
 \end{aligned}$$

Difference Equation:

$$y[n] = x[n] + x[n-3]$$

(b) When $x[n] = \delta[n]$, $y[n] = \delta[n] + \delta[n-3]$



← This is $h[n]$, the impulse response

(c) Need to find where $H(\hat{\omega}) = 0$.

$$1 + e^{-j3\hat{\omega}} = 0$$

$$e^{-j3\hat{\omega}} = -1 = e^{j\pi} e^{j2\pi l}$$

$$\Rightarrow e^{j\hat{\omega}} = e^{-j\pi/3} e^{-j2\pi l/3}$$

$$\Rightarrow \hat{\omega} = -\frac{\pi}{3} - \frac{2\pi}{3}l \quad l=0, 1, 2.$$

$$\hat{\omega} = -\frac{\pi}{3}, -\pi, \text{ and } -\frac{5\pi}{3} \leftarrow \text{same as } +\frac{\pi}{3}$$

$$y[n] = H(\hat{\omega}) A e^{j4} e^{j\hat{\omega}n}$$

Thus when $H(\hat{\omega}) = 0$, the output is zero.



PROBLEM 6.9:

$$\begin{aligned}
 (a) \quad \mathcal{H}(\hat{\omega}) &= (1 - e^{-j\hat{\omega}}) \left(1 - 2(0.5) \cos \frac{\pi}{6} e^{-j\hat{\omega}} + (0.5)^2 e^{-j2\hat{\omega}} \right) \\
 &= (1 - e^{-j\hat{\omega}}) \left(1 - \frac{\sqrt{3}}{2} e^{-j\hat{\omega}} + \frac{1}{4} e^{-j2\hat{\omega}} \right) \\
 &= 1 - \underbrace{\frac{1}{2}(\sqrt{3}+2)}_{-1.866} e^{-j\hat{\omega}} + \underbrace{\left(\frac{1}{4} + \frac{\sqrt{3}}{2}\right)}_{1.116} e^{-j2\hat{\omega}} - \frac{1}{4} e^{-j3\hat{\omega}}
 \end{aligned}$$

Difference Equation:

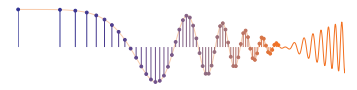
$$y[n] = x[n] - 1.866x[n-1] + 1.116x[n-2] - \frac{1}{4}x[n-3]$$

(b) When $x[n] = \delta[n]$, $y[n] = h[n]$ impulse response

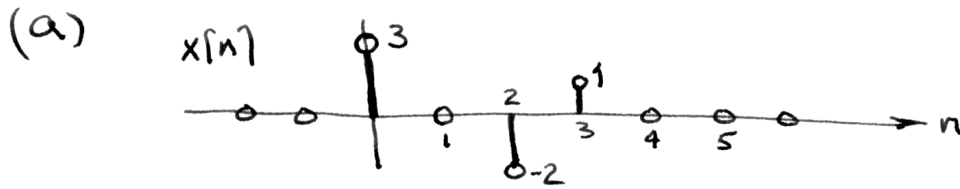
$$h[n] = \delta[n] - 1.866\delta[n-1] + 1.116\delta[n-2] - \frac{1}{4}\delta[n-3]$$

(c) Find $\hat{\omega}$ where $\mathcal{H}(\hat{\omega}) = 0$

The only frequency is $\hat{\omega} = 0$, because then the factor $(1 - e^{-j\hat{\omega}}) = 0$. The other two factors in $\mathcal{H}(\hat{\omega})$ are never zero for $-\pi \leq \hat{\omega} \leq \pi$.



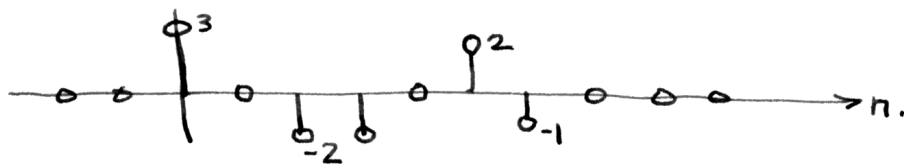
PROBLEM 6.10:



(b) Use LINEARITY & TIME-INVARIANCE

$$\begin{array}{l}
 3\delta[n] \longrightarrow 3\delta[n] - 3\delta[n-3] \\
 -2\delta[n-2] \longrightarrow -2\delta[n-2] + 2\delta[n-5] \\
 \delta[n-3] \longrightarrow \delta[n-3] - \delta[n-6]
 \end{array}
 \left. \vphantom{\begin{array}{l} 3\delta[n] \\ -2\delta[n-2] \\ \delta[n-3] \end{array}} \right\} \begin{array}{l} \text{Add} \\ \text{these} \\ \text{together} \end{array}$$

$$\text{OUTPUT} = 3\delta[n] - 2\delta[n-2] - 2\delta[n-3] + 2\delta[n-5] - \delta[n-6].$$



(c) Use the third input/output pair:

$$\mathcal{X}(\pi/3) = 2 \quad (\text{no phase}).$$

$$\therefore \cos(\pi(n-3)/3) \longrightarrow 2 \cos(\pi(n-3)/3).$$

(d) There is no direct evidence about $\mathcal{X}(\pi/2)$.

BUT use impulse response to get $\{b_k\}$.

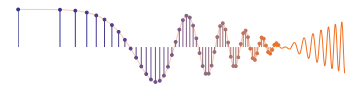
$$h[n] = \delta[n] - \delta[n-3].$$

$$\Rightarrow \{b_k\} = \{1, 0, 0, -1\}$$

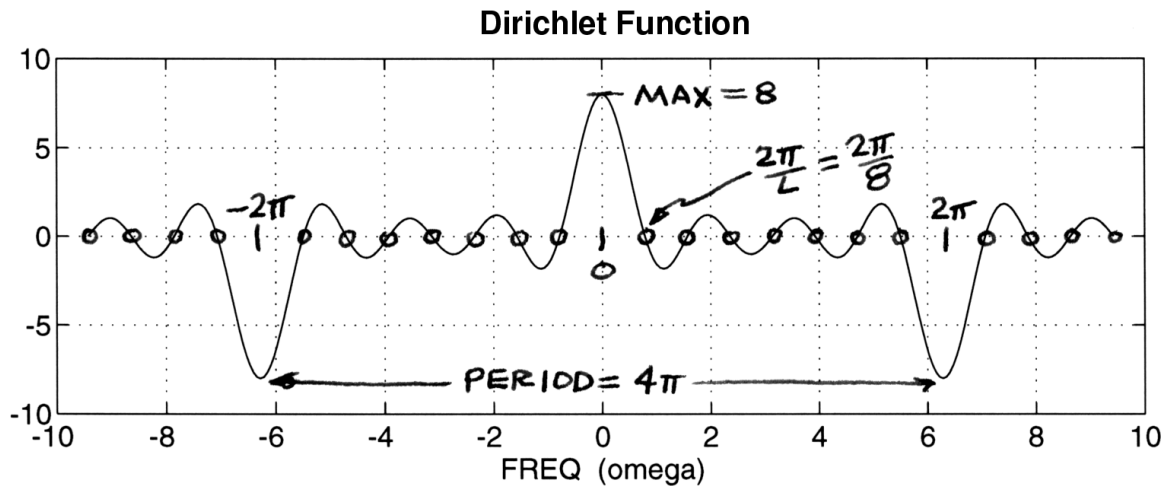
$$\Rightarrow \mathcal{X}(\hat{\omega}) = 1 - e^{-j3\hat{\omega}}$$

$$\therefore \mathcal{X}(\pi/2) \neq 0$$

$$\begin{aligned}
 \mathcal{X}(\pi/2) &= 1 - e^{-j3\pi/2} \\
 &= 1 + j
 \end{aligned}$$



PROBLEM 6.11:



(a) zero crossings at multiples of $\frac{2\pi}{L} = \frac{2\pi}{8} = \frac{\pi}{4}$

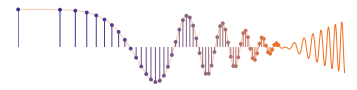
(b) PERIOD = 4π (NOTE: PERIOD = 2π when L is odd)

(c) MAX at $\hat{\omega} = 0, \pm 2\pi, \pm 4\pi$.

Take LIMIT

$$\lim_{\hat{\omega} \rightarrow 0} \frac{\sin 4\hat{\omega}}{\sin \hat{\omega}/2} \rightarrow \frac{4\hat{\omega}}{\hat{\omega}/2} \rightarrow 8$$

BECAUSE $\sin \theta \approx \theta$ when $\theta \rightarrow 0$.



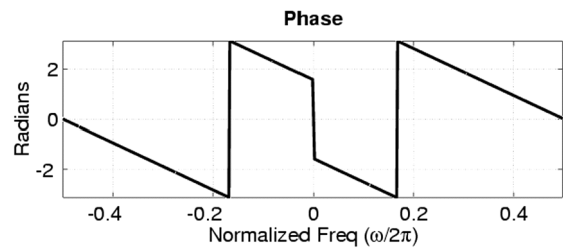
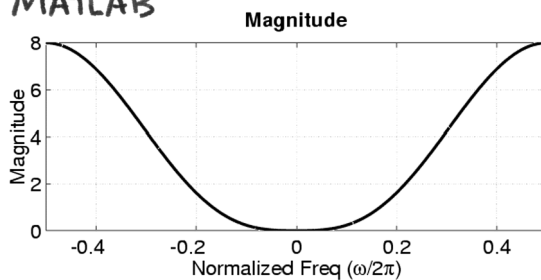
PROBLEM 6.12:

$$y[n] = x[n] - 3x[n-1] + 3x[n-2] - x[n-3]$$

(a) use filter coeffs: $\{b_k\} = \{1, -3, 3, -1\}$

$$\begin{aligned} H(\hat{\omega}) &= 1 - 3e^{-j\hat{\omega}} + 3e^{-j2\hat{\omega}} - e^{-j3\hat{\omega}} = (1 - e^{-j\hat{\omega}})^3 \\ &= e^{-j3\hat{\omega}/2} \left(\frac{e^{j\hat{\omega}/2} - e^{-j\hat{\omega}/2}}{(2j)^3} \right)^3 (2j)^3 \quad \leftarrow (2e^{j\pi/2})^3 = 8e^{-j\pi/2} \\ &= 8e^{j(-\pi/2 - 3\hat{\omega}/2)} \sin^3(\hat{\omega}/2) \end{aligned}$$

(b) MATLAB



(c) $x[n] = 10 + 4 \cos(\frac{\pi}{2}n + \frac{\pi}{4})$

$$y[n] = 10H(0) + 4|H(\frac{\pi}{2})| \cos(\frac{\pi}{2}n + \frac{\pi}{4} + \angle H(\frac{\pi}{2}))$$

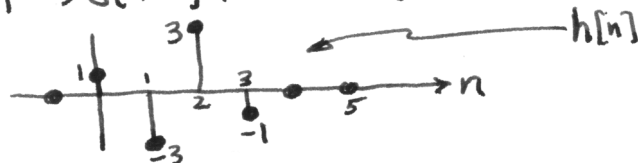
$$\begin{aligned} H(\frac{\pi}{2}) &= 8e^{-j(\pi/2 + 3\pi/4)} \sin^3(\frac{\pi}{4}) \\ &= 8(\frac{\sqrt{2}}{2})^3 e^{-j5\pi/4} = 2\sqrt{2} e^{-j5\pi/4} \end{aligned}$$

$$H(0) = 0$$

$$\begin{aligned} \Rightarrow y[n] &= 10(0) + 8\sqrt{2} \cos(\frac{\pi}{2}n + \frac{\pi}{4} - \frac{5\pi}{4}) \\ &= 8\sqrt{2} \cos(\frac{\pi}{2}n - \pi) \end{aligned}$$

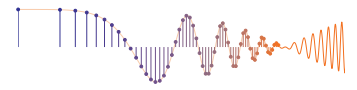
(d) $x[n] = \delta[n]$

$$y[n] = \delta[n] - 3\delta[n-1] + 3\delta[n-2] - \delta[n-3]$$



(e) Use superposition, so just add the results from (c) and (d)

$$y[n] = 8\sqrt{2} \cos(\frac{\pi}{2}n - \pi) + \delta[n] - 3\delta[n-1] + 3\delta[n-2] - \delta[n-3]$$



PROBLEM 6.13:

$$\begin{aligned}
 (a) \quad y[n] &= y_3[n] = x_3[n-1] + x_3[n-2] && x_3[n] = y_2[n] \\
 &= y_2[n-1] + y_2[n-2] \\
 &= (x_2[n-1] + x_2[n-3]) + (x_2[n-2] + x_2[n-4])
 \end{aligned}$$

Now replace $x_2[n]$ with $y_1[n]$

$$\begin{aligned}
 y[n] &= y_1[n-1] + y_1[n-2] + y_1[n-3] + y_1[n-4] \\
 &= (x_1[n-1] - x_1[n-2]) + (x_1[n-2] - x_1[n-3]) \\
 &\quad + (x_1[n-3] - x_1[n-4]) + (x_1[n-4] - x_1[n-5])
 \end{aligned}$$

← CANCEL

$$y[n] = x_1[n-1] - x_1[n-5] \quad x_1[n] = x[n]$$

$$y[n] = x[n-1] - x[n-5]$$

(b) Same thing as part (a) but use $\mathcal{H}_i(\hat{\omega})$

$$\left. \begin{aligned}
 \mathcal{H}_1(\hat{\omega}) &= 1 - e^{-j\hat{\omega}} \\
 \mathcal{H}_2(\hat{\omega}) &= 1 + e^{-j2\hat{\omega}} \\
 \mathcal{H}_3(\hat{\omega}) &= e^{-j\hat{\omega}} + e^{-j2\hat{\omega}}
 \end{aligned} \right\} \text{Multiply these together}$$

$$\begin{aligned}
 \mathcal{H}_6(\hat{\omega}) &= \mathcal{H}_1(\hat{\omega})\mathcal{H}_2(\hat{\omega})\mathcal{H}_3(\hat{\omega}) \\
 &= (1 - e^{-j\hat{\omega}})(1 + e^{-j2\hat{\omega}})(e^{-j\hat{\omega}} + e^{-j2\hat{\omega}}) \\
 &= (1 - e^{-j\hat{\omega}} + e^{-j2\hat{\omega}} - e^{-j3\hat{\omega}})(e^{-j\hat{\omega}} + e^{-j2\hat{\omega}}) \\
 &= e^{-j\hat{\omega}} - e^{-j2\hat{\omega}} + e^{-j3\hat{\omega}} - e^{-j4\hat{\omega}} \\
 &\quad + e^{-j2\hat{\omega}} - e^{-j3\hat{\omega}} + e^{-j4\hat{\omega}} - e^{-j5\hat{\omega}}
 \end{aligned}$$

$$\mathcal{H}_6(\hat{\omega}) = e^{-j\hat{\omega}} - e^{-j5\hat{\omega}}$$

$$\Rightarrow y[n] = x[n-1] - x[n-5]$$

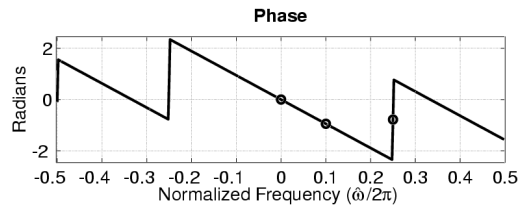
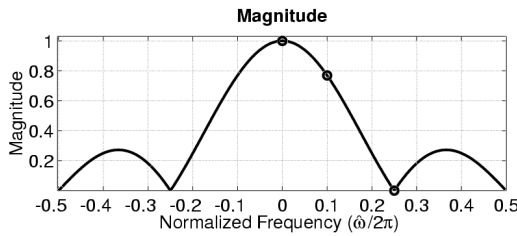


PROBLEM 6.14:

(a) $h[n] = \frac{1}{4} \delta[n] + \frac{1}{4} \delta[n-1] + \frac{1}{4} \delta[n-2] + \frac{1}{4} \delta[n-3]$

(b)
$$\begin{aligned} \mathcal{H}(\hat{\omega}) &= \frac{1}{4} (1 + e^{-j\hat{\omega}} + e^{-j2\hat{\omega}} + e^{-j3\hat{\omega}}) \\ &= \frac{1}{4} e^{-j\frac{3}{2}\hat{\omega}} (e^{j\frac{3}{2}\hat{\omega}} + e^{j\frac{1}{2}\hat{\omega}} + e^{-j\frac{1}{2}\hat{\omega}} + e^{-j\frac{3}{2}\hat{\omega}}) \\ &= \frac{1}{2} e^{-j\frac{3}{2}\hat{\omega}} (\cos(\frac{3}{2}\hat{\omega}) + \cos(\frac{1}{2}\hat{\omega})) \end{aligned}$$

(c) MATLAB



(d) $x[n] = 5 + 4 \cos(0.2\pi n) + 3 \cos(0.5\pi n + \pi/4)$
 Need $\mathcal{H}(0)$ Need $\mathcal{H}(0.2\pi)$ $\mathcal{H}(0.5\pi)$

$\mathcal{H}(0) = 1$

$\mathcal{H}(0.2\pi) = 0.769 e^{-j0.3\pi}$ ANGLE = $-54^\circ = -0.942 \text{ rad}$

$\mathcal{H}(0.5\pi) = 0$

$\Rightarrow y[n] = 5 + \underbrace{4(0.769)}_{3.078} \cos(0.2\pi n - 0.3\pi)$

(e) $x_1[n] = 0$ for $n < 0$

$y_1[n] = \frac{1}{4} (x_1[n] + x_1[n-1] + x_1[n-2] + x_1[n-3])$

Since $x_1[n] = x[n]$ for $n \geq 0$, the filtered outputs will be the same when $n-3 \geq 0$

$\Rightarrow n \geq 3$ is the region where $y_1[n] = y[n]$

Here's a table of the first few values:

n	-1	0	1	2	3	4 ...
$y[n]$	4.029	6.809	7.927	7.927	6.809	5 ...
$y_1[n]$	0	2.780	4.309	5.338	6.809	5 ...



PROBLEM 6.15:

$$x(t) = 10 + 8 \cos(200\pi t) + 6 \cos(500\pi t + \pi/4)$$

$$f_s = 1000$$

$$x[n] = x(t) \Big|_{t=n/f_s = n/1000}$$

$$= 10 + 8 \cos\left(200 \frac{\pi n}{1000}\right) + 6 \cos\left(\frac{500\pi n}{1000} + \pi/4\right)$$

$$= 10 + 8 \cos(0.2\pi n) + 6 \cos(0.5\pi n + \pi/4)$$

\nearrow Need $H(0)$ \nearrow Need $H(0.2\pi)$ \nearrow $H(0.5\pi) = 0$

Use frequency response values from Prob. 6.14

$$H(0) = 1 \quad H(0.2\pi) = 0.769 e^{-j0.3\pi}$$

$$y[n] = 10 + (0.769) 8 \cos(0.2\pi n - 0.3\pi)$$

$$= 10 + 6.156 \cos(0.2\pi n - 0.3\pi)$$

$$y(t) = y[n] \Big|_{n \leftarrow f_s t = 1000t}$$

$$= 10 + 6.156 \cos(200\pi t - 0.3\pi)$$



PROBLEM 6.16:

(a) System 2 will block the DC component which is a constant. $\mathcal{H}_2(0) = 0$. Also $y[n] = v[n] - v[n-1]$ so the differencing operator removes DC.

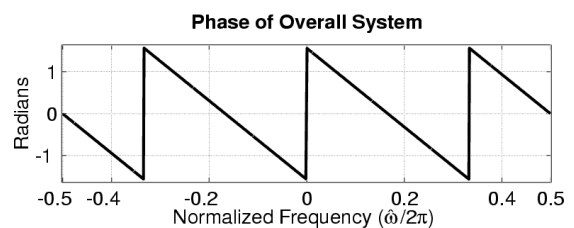
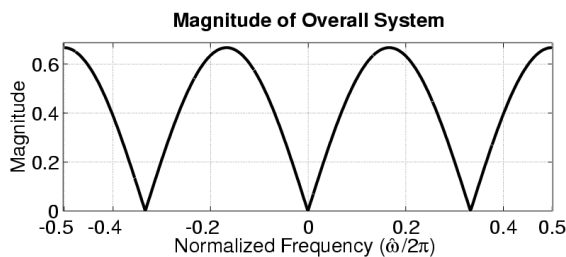
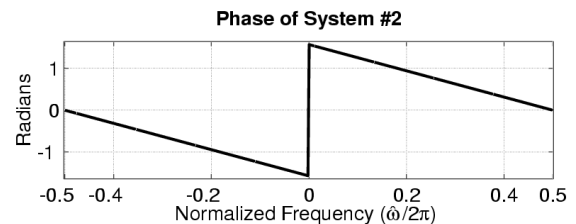
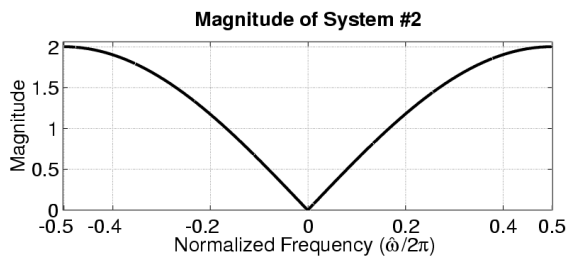
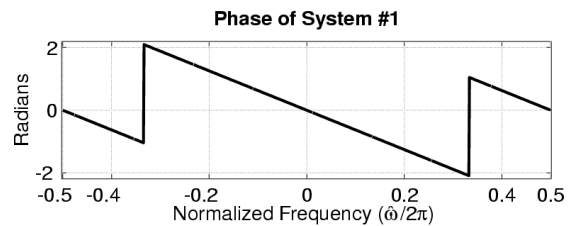
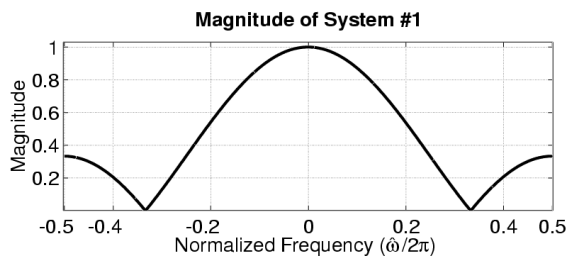
(b) $\mathcal{H}(\hat{\omega}) = \mathcal{H}_1(\hat{\omega}) \mathcal{H}_2(\hat{\omega})$

$\mathcal{H}_1(\hat{\omega}) = \frac{1}{3} + \frac{1}{3}e^{j\hat{\omega}} + \frac{1}{3}e^{j2\hat{\omega}}$

$\mathcal{H}_2(\hat{\omega}) = 1 - e^{-j\hat{\omega}}$

$\mathcal{H}(\hat{\omega}) = \frac{1}{3} - \frac{1}{3}e^{-j3\hat{\omega}}$

(c) MATLAB



(d) $y[n] = \frac{1}{3}x[n] - \frac{1}{3}x[n-3]$



PROBLEM 6.17:

A linear time-invariant system is described by the difference equation

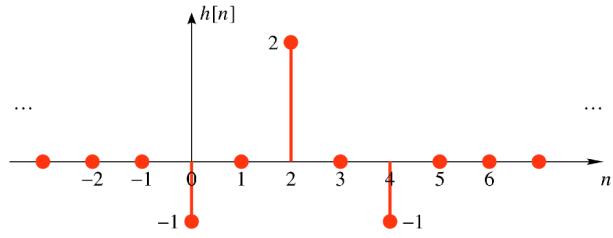
$$y[n] = -x[n] + 2x[n - 2] - x[n - 4]$$

- (a) Find the impulse response $h[n]$ and plot it.

Solution:

Let $x[n]$ be $\delta[n]$, and find the output:

$$h[n] = -\delta[n] + 2\delta[n - 2] - \delta[n - 4]$$



- (b) Determine an equation for the frequency response $H(e^{j\hat{\omega}})$ and express it as $H(e^{j\hat{\omega}}) = e^{-j\hat{\omega}n_0} R(e^{j\hat{\omega}})$, where $R(e^{j\hat{\omega}})$ is a real function and n_0 is an integer.

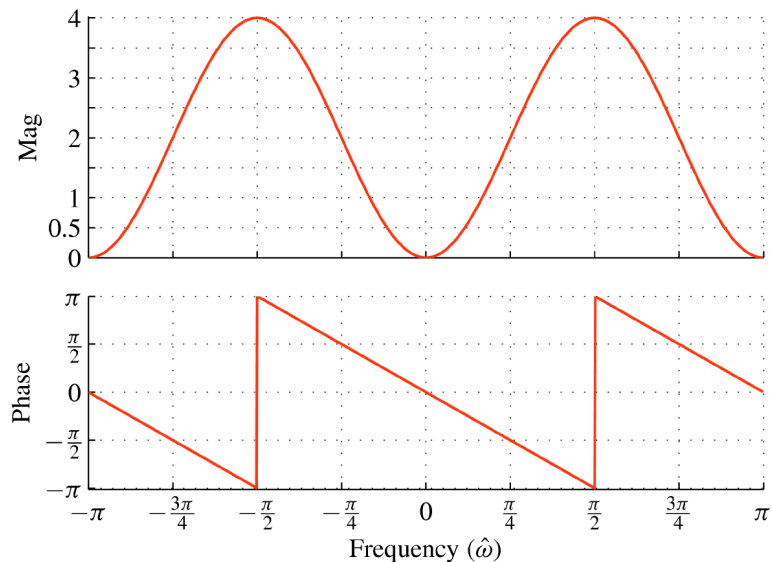
Solution: Plug into the frequency response formula:

$$\begin{aligned} H(e^{j\hat{\omega}}) &= -e^{j0} + 2e^{-j2\hat{\omega}} - e^{-j4\hat{\omega}} \\ &= e^{-j2\hat{\omega}} \left(-e^{j2\hat{\omega}} + 2 - e^{-j2\hat{\omega}} \right) \\ &= e^{-j2\hat{\omega}} (2 - 2\cos(2\hat{\omega})) = e^{-j\hat{\omega}n_0} R(e^{j\hat{\omega}}) \end{aligned}$$

Thus, $n_0 = 2$ and $R(e^{j\hat{\omega}}) = 2 - 2\cos(2\hat{\omega})$.

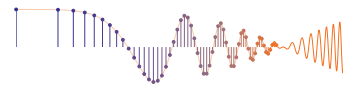
- (c) Carefully sketch and label a plot of $|H(e^{j\hat{\omega}})|$ for $-\pi < \hat{\omega} < \pi$.

Solution: Since $R(e^{j\hat{\omega}}) \geq 0$, the magnitude is $|H(e^{j\hat{\omega}})| = R(e^{j\hat{\omega}}) = 2 - 2\cos(2\hat{\omega})$.



- (d) Carefully sketch and label a plot of the principal value of the $\angle H(e^{j\hat{\omega}})$ for $-\pi < \hat{\omega} < \pi$.

Solution: The phase is $\angle H(e^{j\hat{\omega}}) = -2\hat{\omega}$, but the principal value wraps at $\hat{\omega} = \pi/2$ when $\angle H(e^{j\hat{\omega}})$ is outside of the range $[-\pi, \pi]$.



PROBLEM 6.18:

$$X[n] = 5 + 20 \cos\left(\frac{\pi}{2}n + \frac{\pi}{4}\right) + 10\delta[n-3]$$

↑ Need $\mathcal{H}(0)$
↙ DEPENDS on $\mathcal{H}(\pi/2)$
↖ Need impulse response $h[n]$

$$\begin{aligned} \mathcal{H}(0) &= (1-j)(1-(-j))(1+1) \\ &= (1-j)(1+j)2 = 2 \cdot 2 = 4 \end{aligned}$$

$$\begin{aligned} \mathcal{H}(\pi/2) &= (1-j e^{-j\pi/2})(1+j e^{-j\pi/2})(1+e^{-j\pi/2}) \\ &= (1-j(-j))(1+j(-j))(1-j) \\ &= (1-1)(1+1)(1-j) = 0 \end{aligned}$$

To find $h[n]$, multiply out $\mathcal{H}(\hat{\omega})$

$$\begin{aligned} \mathcal{H}(\hat{\omega}) &= (1-j e^{-j\hat{\omega}} + j e^{-j\hat{\omega}} + e^{-j2\hat{\omega}})(1+e^{-j\hat{\omega}}) \\ &= (1+e^{-j2\hat{\omega}})(1+e^{-j\hat{\omega}}) \\ &= 1 + e^{-j\hat{\omega}} + e^{-j2\hat{\omega}} + e^{-j3\hat{\omega}} \end{aligned}$$

$$\Rightarrow h[n] = \delta[n] + \delta[n-1] + \delta[n-2] + \delta[n-3]$$

Finally,

$$\begin{aligned} y[n] &= 5(4) + 0 + 10h[n-3] \\ &= 20 + 10\delta[n-3] + 10\delta[n-4] + 10\delta[n-5] + 10\delta[n-6] \end{aligned}$$



PROBLEM 6.19:

$$(a) \mathcal{H}(\hat{\omega}) = \mathcal{H}_1(\hat{\omega}) \mathcal{H}_2(\hat{\omega})$$

$$\mathcal{H}_2(\hat{\omega}) = 1 - e^{-j\hat{\omega}} + e^{-j2\hat{\omega}} - e^{-j3\hat{\omega}}$$

$$\mathcal{H}_1(\hat{\omega}) = 1 + 2e^{j\hat{\omega}} + e^{j2\hat{\omega}}$$

Multiply:

$$\mathcal{H}(\hat{\omega}) = 1 - e^{-j\hat{\omega}} + e^{-j2\hat{\omega}} - e^{-j3\hat{\omega}} + 2e^{-j\hat{\omega}} - 2e^{-j2\hat{\omega}} + 2e^{-j3\hat{\omega}} - 2e^{-j4\hat{\omega}} + e^{j2\hat{\omega}} - e^{-j3\hat{\omega}} + e^{-j4\hat{\omega}} - e^{-j5\hat{\omega}}$$

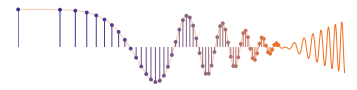
$$\mathcal{H}(\hat{\omega}) = 1 + e^{-j\hat{\omega}} - e^{-j4\hat{\omega}} - e^{-j5\hat{\omega}}$$

MANY TERMS
CANCEL OUT

$$(b) h[n] = \delta[n] + \delta[n-1] - \delta[n-4] - \delta[n-5]$$

(c) The polynomial coefficients of $\mathcal{H}(\hat{\omega})$ define $\{b_k\}$ as $\{1, 1, 0, 0, -1, -1\}$. Use $\{b_k\}$ as filter coefficients:

$$y[n] = x[n] + x[n-1] - x[n-4] - x[n-5]$$



PROBLEM 6.20:

(a) The highest frequency in $x(t)$ is $\omega_0 = 2\pi(500)$

To avoid aliasing we must sample at $f_s > 2f_{MAX}$

$$\Rightarrow f_s > 2(500\text{Hz}) = 1000 \text{ samples/sec.}$$

(b) $h[n] = \delta[n-10]$, f_s and ω_0 to be determined

$$x[n] = 10 + 20 \cos(\omega_0 n / f_s + \pi/3)$$

$$y[n] = x[n-10] = 10 + 20 \cos(\omega_0 \frac{(n-10)}{f_s} + \pi/3)$$

$$y(t) = y[n] \Big|_{n \leftarrow f_s t} = 10 + 20 \cos(\frac{\omega_0}{f_s} (f_s t - 10) + \pi/3)$$

Since we want $y(t) = x(t - 0.001)$, we need

$$\frac{\omega_0}{f_s} (10) = (0.001) \omega_0$$

$$\Rightarrow 10/f_s = 1/1000 \Rightarrow f_s = 10,000 \text{ Hz}$$

In order for the output frequency to be the same as the input frequency ω_0 , there must be no aliasing. $\Rightarrow 2\omega_0 < 2\pi f_s$

$$\Rightarrow \omega_0 < 2\pi(500) \text{ rad/sec}$$

(c) To have $y(t) = A$, we need $y[n] = \text{constant}$.

Since $x[n] = 10 + 20 \cos(\omega_0 n / f_s + \pi/3)$ $f_s = 2000$

the filter must "null out" the cosine term

$$\Rightarrow \frac{\omega_0}{f_s} = \hat{\omega}_{NULL} \text{ where } \hat{\omega}_{NULL} \text{ is one of the zeros of } \mathcal{H}(\hat{\omega})$$

$$\mathcal{H}(\hat{\omega}) = 0 \text{ when } \hat{\omega} = 2\pi/5, 4\pi/5, -2\pi/5, -4\pi/5$$

$$\therefore \omega_0 = f_s \hat{\omega}_{NULL} = \{ 2\pi(400), 2\pi(800), -2\pi(400), -2\pi(800) \}$$

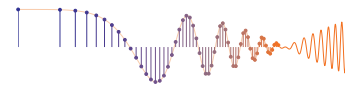
We must include all aliases:

$$2\pi(400), 2\pi(2400), 2\pi(4400), \dots \quad 2\pi(400 + 2000l)$$

$$2\pi(800), 2\pi(2800), 2\pi(4800), \dots \quad 2\pi(800 + 2000l)$$

$$2\pi(-400), 2\pi(1600), 2\pi(3600), \dots \quad 2\pi(-400 + 2000l)$$

$$2\pi(-800), 2\pi(1200), 2\pi(3200), \dots \quad 2\pi(-800 + 2000l)$$



PROBLEM 6.21:

$$(a) \quad x[n] = 10 + 10 \cos(0.2\pi n) + 10 \cos(0.5\pi n)$$

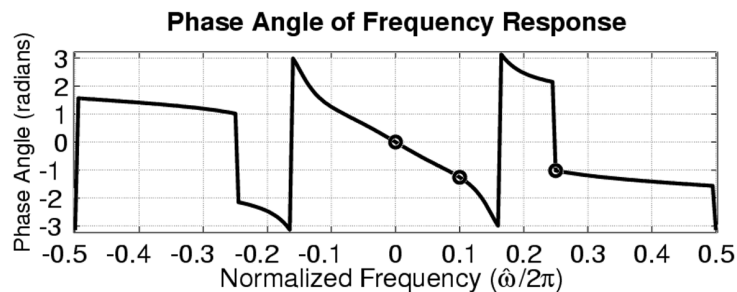
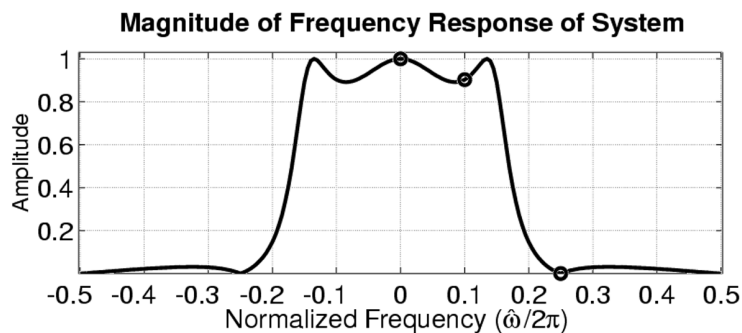
Need $\mathcal{H}(0)$
Need $\mathcal{H}(0.2\pi)$
Need $\mathcal{H}(0.5\pi)$

$$\mathcal{H}(0) = 1$$

$$\mathcal{H}(0.2\pi) = 0.9027 e^{-j0.4\pi} \quad \text{ANGLE} = -71.98^\circ = -1.26 \text{ rads}$$

$$\mathcal{H}(0.5\pi) = 0.00089 e^{-j0.323\pi} \quad \text{ANGLE} = -58.22^\circ = -1.02 \text{ rads}$$

$$y[n] = 10 + 9.027 \cos(0.2\pi n - 0.4\pi) + \underbrace{(0.00089) 10 \cos(0.5\pi n - 0.323\pi)}_{\text{Very close to zero}}$$



(b) The discontinuity at $\hat{\omega} = 2\pi(0.25)$ is caused by the zero near $\hat{\omega} = 2\pi(0.25)$. There is a sign change in $\mathcal{H}(\hat{\omega})$ which means the phase changes by π . The discontinuity at $\hat{\omega} = 2\pi(0.17)$ is a " 2π -jump" which happens when the principal value of the phase tries to cross π . The arctangent calculation flips the value from π to $-\pi$ creating a " 2π jump."